EFFICIENT IMAGE RECONSTRUCTION FOR GIGAPIXEL QUANTUM IMAGE SENSORS

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ABSTRACT

Recent advances in materials, devices and fabrication technologies have motivated a strong momentum in developing solid-state sensors that can detect individual photons in space and time. It has been envisioned that such sensors can eventually achieve very high spatial resolutions (e.g., 10^9 pixels/chip) as well as high frame rates (e.g., 10⁶ frames/sec). In this paper, we present an efficient algorithm to reconstruct images from the massive binary bit-streams generated by these sensors. Based on the concept of alternating direction method of multipliers (ADMM), we transform the computationally intensive optimization problem into a sequence of subproblems, each of which has efficient implementations in the form of polyphase-domain filtering or pixel-wise nonlinear mappings. Moreover, we reformulate the original maximum likelihood estimation as maximum a posterior estimation by introducing a total variation prior. Numerical results demonstrate the strong performance of the proposed method, which achieves several dB's of improvement in PSNR and requires a shorter runtime as compared to standard gradient-based approaches.

Index Terms— Image reconstruction, quantum image sensors, gigapixel imaging, ADMM

1. INTRODUCTION

Thanks to recent advances in materials and fabrication technologies, there has been an emerging class of solid-state imaging sensors that are capable of achieving single-photon sensitivity, sub-nanosecond time resolution and rapidly increasing spatial resolutions [1, 2]. These devices, collectively referred to as quantum image sensors (QIS) in this work, are envisioned as the next generation imaging technology after CMOS [3], with numerous applications [4–7].

The operational principle of a QIS is analogous to photographic film: Photon flux reaching a pixel of the single-photon detector triggers a binary response, thus generates a 1-bit signal revealing the intensity of the flux during the exposure. With high frame rates (e.g., 10^6 frames/sec) and increasing spatial resolutions (e.g., 10^9 pixels/chip as envisioned in [3]), the QIS generates a massive stream of bits, which must be "decoded" to recover the underlying image.

The 1-bit measurements acquired by the QIS follow a quantized Poisson process. Therefore, image reconstruction from these measurements can be solved via maximum likelihood estimation (MLE). However, except for simple cases where the photon flux is a piecewise constant function, the MLE does not have a closed-form solution in general, even though the problem is convex [8]. Iterative algorithms have been proposed, e.g., [9–11], but how to improve the slow convergence is still an open problem.

By recognizing the separable structure of the MLE's objective function, we propose a new algorithm based on the concept of alternating direction method of multipliers (ADMM) [12, 13]. The contributions of this paper are two-fold. First, we present fast solutions to the subproblems associated with the ADMM formulation. One key technique we use is the polyphase representation, a powerful tool from multirate signal processing [14]. Second, we introduce a total variation (TV) prior and modify the MLE problem to a maximum a posterior (MAP) problem. We show empirically that the MAP formulation significantly improves the reconstruction quality with only marginal increase in computational cost.

The rest of the paper is organized as follows. We first present the modeling of QIS and elaborate on the MLE problem in Section 2. In Section 3 we discuss the proposed ADMM algorithm. Experimental results are shown in Section 4 and we conclude in Section 5.

2. MODEL AND PROBLEM FORMULATION

The signal processing model of a QIS sensor consists of two major components: (1) The light exposure stage which models the propagation of light from scene to sensor; (ii) The sensing stage which models the conversion of the light field to the binary measurements. A pictorial illustration is shown in Figure 1, and each component is described as follows.

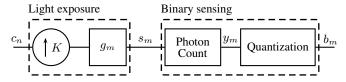


Fig. 1: The two mathematical components of modeling the image formation of QIS (c.f. [10]).

2.1. Modeling of Light Exposure

Consider a light field intensity $\lambda(x)$ of spatial coordinate x. Without loss of generality we shall assume that $\lambda(x)$ is a one-dimensional function and that the support of the coordinate is the unit interval, i.e., $0 \le x \le 1$. Furthermore, due to the presence of diffraction limit of the lens, we model $\lambda(x)$ as a function in the shift-invariant space spanned by a non-negative interpolation kernel $\varphi(x)$:

$$\lambda(x) = \frac{N}{\tau} \sum_{n=0}^{N-1} c_n \varphi(Nx - n), \tag{1}$$

where τ is the exposure time, $\{c_n : c_n \geq 0\}$ is a set of variables encoding the scene, and N is the number of variables. The constant N/τ is not essential, but it simplifies the analysis below.

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Suppose that the sensor consists of M pixels per unit space and that the mth pixel covers the area [m/M, (m+1)/M], then the total light exposure accumulated on the surface area of the mth pixel within a time period $[0,\tau]$ is

$$s_m \stackrel{\text{def}}{=} \int_0^{\tau} \int_{m/M}^{(m+1)/M} \lambda(x) dx dt = \tau \langle \lambda(x), \beta(Mx - m) \rangle, \quad (2)$$

where $\langle \cdot, \cdot \rangle$ represents the standard L^2 inner product, and $\beta(x)$ is the box function defined as $\beta(x) = 1$ for $0 \le x \le 1$ and is zero otherwise. Substituting (1) into (2) yields

$$s_m = \sum_{n=0}^{N-1} c_n g_{m-Kn},\tag{3}$$

where $K \stackrel{\text{def}}{=} M/N$ is the spatial *oversampling ratio*, and

$$g_m \stackrel{\text{def}}{=} \langle \varphi(x), \beta(Kx - m) \rangle$$
 (4)

is the mth coefficient of a discrete filter. Using matrix-vector notations, the expression in (3) can be equivalently written as s = Gc, for appropriate $s \in \mathbb{R}^{M \times 1}$, $c \in \mathbb{R}^{N \times 1}$ and $G \in \mathbb{R}^{M \times N}$.

2.2. Modeling of Binary Sensing

Let y_m be the number of photons impinging on the surface of the mth pixel during an exposure period $[0, \tau]$. We model the relationship between s_m and y_m as a Poisson process Y_m :

$$\mathbb{P}[Y_m = y_m; s_m] = \frac{s_m^{y_m} e^{-s_m}}{y_m!}.$$
 (5)

The binary sensing of the QIS quantizes the observed photon count y_m through a threshold operation. In this paper, we assume that the threshold level is set to unity so that the binary response is

$$B_m = 1 \text{ if } Y_m \ge 1 \quad \text{and} \quad B_m = 0 \text{ if } Y_m = 0.$$
 (6)

Hence, the probability of observing B_m is

$$\mathbb{P}[B_m = 1] = 1 - e^{-s_m} \quad \text{and} \quad \mathbb{P}[B_m = 0] = e^{-s_m}.$$
 (7)

2.3. Maximum Likelihood Estimation

The image reconstruction task of the QIS can be formulated as the following maximum likelihood estimation (MLE):

$$\boldsymbol{c}^* = \underset{\boldsymbol{s} = \boldsymbol{G}\boldsymbol{c}}{\operatorname{arg\,max}} \quad \prod_{m=1}^M \mathbb{P}[B_m = b_m; s_m]$$
 (8)

$$= \underset{s=Gc}{\text{arg min}} - \sum_{m=1}^{M} \log \left((1 - b_m) e^{-s_m} + b_m (1 - e^{-s_m}) \right).$$

Defining $F(s) \stackrel{\text{def}}{=} -\sum_{m=1}^{M} \log((1-b_m)e^{-s_m} + b_m(1-e^{-s_m}))$ and introducing a prior p(c), we can further modify the MLE to a maximum a posterior (MAP) estimation problem:

$$c^* = \underset{s = Gc}{\operatorname{arg \, min}} F(s) - \log p(c).$$
 (9)

In this paper, we consider the total variation prior

$$p(\mathbf{c}) \propto \exp(-\lambda \|\mathbf{c}\|_{TV}),$$
 (10)

where $\|\boldsymbol{c}\|_{TV} = \|\boldsymbol{D}\boldsymbol{c}\|_1$ and \boldsymbol{D} is the first order finite difference operator. Thus, (9) is equivalent to

minimize
$$F(s) + \lambda ||Dc||_1$$
, subject to $s = Gc$, (11)

which is a convex optimization problem. In the following sections we present an efficient algorithm to solve (11).

3. IMAGE RECONSTRUCTION ALGORITHM

The proposed image reconstruction algorithm is based on the concept of alternating direction method of multipliers (ADMM) [13]. The idea is to consider the following equivalent constrained problem

minimize
$$F(s) + \lambda ||v||_1$$
, subject to $s = Gc$, $v = Dc$, (12)

by defining an auxiliary variable $v \in \mathbb{R}^{N \times 1}$. Then, we consider the augmented Lagrangian function:

$$\mathcal{L}(\boldsymbol{c}, \boldsymbol{s}, \boldsymbol{v}, \boldsymbol{z}, \boldsymbol{r}) \stackrel{\text{def}}{=} F(\boldsymbol{s}) + \lambda \|\boldsymbol{v}\|_1 - \boldsymbol{z}^T (\boldsymbol{s} - \boldsymbol{G}\boldsymbol{c}) + \frac{\rho}{2} \|\boldsymbol{s} - \boldsymbol{G}\boldsymbol{c}\|^2 - \boldsymbol{r}^T (\boldsymbol{v} - \boldsymbol{D}\boldsymbol{c}) + \frac{\gamma}{2} \|\boldsymbol{v} - \boldsymbol{D}\boldsymbol{c}\|^2, \quad (13)$$

where $z \in \mathbb{R}^{M \times 1}$ and $r \in \mathbb{R}^{N \times 1}$ are the Lagrange multipliers associated with the constraints s = Gc and v = Dc, respectively, and $\rho > 0$ and $\gamma > 0$ are parameters that control the emphasis of the quadratic penalty functions. ADMM states that the saddle point of \mathcal{L} is the optimal solution of (12). Consequently, (12) can be solved by alternatingly solving the following sequence of subproblems:

$$c^{(k+1)} = \underset{c}{\operatorname{arg \, min}} \ \mathcal{L}(c, s^{(k)}, v^{(k)}, z^{(k)}, r^{(k)}),$$
 (14a)

$$s^{(k+1)} = \underset{\boldsymbol{z}}{\operatorname{arg \, min}} \ \mathcal{L}(\boldsymbol{c}^{(k+1)}, s, \boldsymbol{v}^{(k)}, \boldsymbol{z}^{(k)}, \boldsymbol{r}^{(k)}),$$
 (14b)

$$v^{(k+1)} = \underset{v}{\operatorname{arg min}} \ \mathcal{L}(c^{(k+1)}, s^{(k+1)}, v, z^{(k)}, r^{(k)}),$$
 (14c)

$$z^{(k+1)} = z^{(k)} - \rho \left(s^{(k+1)} - Gc^{(k+1)} \right),$$
 (14d)

$$r^{(k+1)} = r^{(k)} - \gamma \left(v^{(k+1)} - Dc^{(k+1)} \right).$$
 (14e)

We now discuss how each of the above subproblems is solved. For notational simplicity we drop the iteration number $(\cdot)^{(k)}$.

3.1. c-Subproblem

The c-subproblem is formulated by dropping terms independent of c in (14a), yielding

minimize
$$z^T G c + r^T D c + \frac{\rho}{2} ||s - G c||^2 + \frac{\gamma}{2} ||v - D c||^2$$
. (15)

Setting the first order derivative to zero, the solution of (15) is

$$\boldsymbol{c} = \mathcal{P}_{\mathcal{D}} \left[(\rho \boldsymbol{G}^{T} \boldsymbol{G} + \gamma \boldsymbol{D}^{T} \boldsymbol{D})^{-1} \left(\boldsymbol{G}^{T} (\rho \boldsymbol{s} - \boldsymbol{z}) + \boldsymbol{D}^{T} (\gamma \boldsymbol{v} - \boldsymbol{r}) \right) \right],$$
(16)

where $\mathcal{P}_{\mathcal{D}}$ is the projection onto the search domain $\mathcal{D} \stackrel{\text{def}}{=} [0,1]^N$. We apply $\mathcal{P}_{\mathcal{D}}$ to ensure that all estimations of c lie in \mathcal{D} .

The challenge of (16) is the matrix inverse. By construction, D^TD is a circulant matrix and hence it is diagonalizable using the discrete Fourier transform (DFT). Thus it remains to show that G^TG is also a circulant matrix, for then the matrix inverse can be implemented efficiently in the Fourier domain.

To show that G^TG is circulant, it suffices to show that G^TG is a convolution matrix representing a finite impulse response (FIR) filter. Expressing the operations of G and G^T explicitly, we note that G is equivalent to applying an up-sampling operation followed by an FIR filter $\{g_m\}$, where as G^T is equivalent to applying the time reversal filter $\{g_{-m}\}$ followed by a down-sampling operation. The concatenation of these operations is the system shown in the middle of Figure 2.

$$\mathbf{G}^T \mathbf{G} \equiv - \begin{pmatrix} & & & \\ & & & \\ & & & \end{pmatrix} K = g_{-m} * g_m$$

Fig. 2: The equivalence between the matrix operation G^TG , the system, and the finite impulse response filter \hat{q}_{Km} .

To simplify the system shown in Figure 2, we denote

$$\widehat{g}_m \stackrel{\text{def}}{=} g_{-m} * g_m \tag{17}$$

and split \widehat{g}_m into K nonoverlapping polyphase components $\widehat{g}_{0,m}$, $\widehat{g}_{1,m},\ldots,\widehat{g}_{K-1,m}$, defined as

$$\widehat{g}_{k,m} = \widehat{g}_{Km+k}, \quad \text{for } 0 \le k < K. \tag{18}$$

Then, by using the *z*-transform and multirate identities [14] one can show the following proposition.

Proposition 1. G^TG is a single convolution matrix whose impulse response is equal to \widehat{g}_{Km} , where \widehat{g}_m is defined in (17).

The implication of Proposition 1 is that the matrix G^TG has a circulant structure given by the FIR filter \widehat{g}_{Km} . This suggests that diagonalizing G^TG is equivalent to applying DFT to $\{\widehat{g}_{Km}\}$, which can be done offline as $\{\widehat{g}_{Km}\}$ is fully specified by the interpolation kernel $\varphi(x)$ through (4) and (17).

3.2. s-Subproblem

Letting d = Gc in (14b) and eliminating terms independent of s, the s-subproblem can be written as

minimize
$$F(s) - z^T s + \frac{\rho}{2} ||s - d||^2$$
. (19)

Since F(s) is separable, we express (19) as

minimize
$$\sum_{m=1}^{M} \left[-\log \left((1 - b_m) e^{-s_m} + b_m (1 - e^{-s_m}) \right) - z_m s_m + \frac{\rho}{2} (s_m - d_m)^2 \right].$$
 (20)

Thus, (20) is minimized if each of individual terms in the sum is minimized. Considering the mth term of (20), we need to solve two cases:

if
$$b_m = 0$$
: $\min_{s_m} s_m - z_m s_m + \frac{\rho}{2} (s_m - d_m)^2$, (21a)

if
$$b_m = 1$$
: $\min_{s_m} -\log(1 - e^{-s_m}) - z_m s_m + \frac{\rho}{2} (s_m - d_m)^2$.

We note that (21a) is a quadratic minimization. Thus it has a unique solution given by

$$s_m = (1/\rho)(1 - z_m) + d_m.$$
 (22)

For (21b), the first order optimality condition implies that

$$\frac{1}{1 - e^{s_m}} - z_m + \rho(s_m - d_m) = 0. \tag{23}$$

However, (23) is a transcendental equation. For arbitrary choices of (z_m, d_m, ρ) , there is no closed form solution in general.

To solve for a numerical solution of (23), we construct a nonlinear mapping (for a fixed ρ) as follows. We first consider the equation

$$\frac{1}{e^x - 1} = \rho x - w, (24)$$

where we identify (24) as (23) by noting $x=s_m$ and $w=z_m+\rho d_m$. Let $[w_{\min},\,w_{\max}]$ be the interval in which w is defined, we partition $[w_{\min},\,w_{\max}]$ into L equidistance subintervals and consider a sequence

$$w_{\ell} = w_{\min} + \ell(\Delta w), \tag{25}$$

for $\ell=1,\ldots,L$, where $\Delta w=(w_{\max}-w_{\min})/L$ is the length of the subinterval. For each ℓ , we numerically determine the solution of (24) using fzero in MATLAB. This returns a sequence $\{x(w_\ell):\ell=1,\ldots,L\}$, which are the solutions of (24) for L particular values of w. For any other $w\in[w_{\min},w_{\max}]$, we interpolate the result using a linear interpolation scheme, given by

$$x(w) = x(w_{\ell}) + \left(\frac{x(w_{\ell+1}) - x(w_{\ell})}{w_{\ell+1} - w_{\ell}}\right) (w - w_{\ell}), \qquad (26)$$

where ℓ is chosen such that $w_{\ell} \leq w \leq w_{\ell+1}$.

We note that the computational cost of solving (23) includes the construction of a sequence $\{x(w_\ell):\ell=1,\ldots,L\}$ and a linear interpolation step. The former can be determined offline because it is independent of the image data. The latter is evaluated using (26), which is a scalar update and can be implemented in parallel.

3.3. v-subproblem

The v-subproblem defined in (14c) is

$$\underset{\boldsymbol{v}}{\text{minimize}} \quad \lambda \|\boldsymbol{v}\|_{1} - \boldsymbol{r}^{T}(\boldsymbol{v} - \boldsymbol{D}\boldsymbol{c}) + \frac{\gamma}{2} \|\boldsymbol{v} - \boldsymbol{D}\boldsymbol{c}\|^{2}.$$
 (27)

Applying the shrinkage formula [12], the solution is

$$v = \max(|Dc + r/\gamma| - \lambda/\gamma, 0) \cdot \text{sign}(Dc + r/\gamma).$$
 (28)

The overall algorithm is summarized in Algorithm 1.

Algorithm 1 Image Reconstruction for QIS

while $\|oldsymbol{c}^{(k+1)} - oldsymbol{c}^{(k)}\|/\|oldsymbol{c}^{(k)}\| \geq \mathsf{tol}\; \mathsf{do}$

• s-subproblem:

for $m = 1, \ldots, M$ do

If
$$b_m = 0$$
, solve $1 - z_m + \rho(s_m - d_m) = 0$.
If $b_m = 1$, solve $1/(1 - e^{s_m}) - z_m + \rho(s_m - d_m) = 0$.

end for

• c-subproblem: solve

$$(\rho \mathbf{G}^{\mathbf{T}} \mathbf{G} + \gamma \mathbf{D}^T \mathbf{D}) \mathbf{c} = \mathbf{G}^T (\rho \mathbf{s} - \mathbf{z}) + \mathbf{D}^T (\gamma \mathbf{v} - \mathbf{r}).$$

• *v*-subproblem:

$$v = \max(|Dc + r/\gamma| - \lambda/\gamma, 0) \cdot \text{sign}(Dc + r/\gamma).$$

- z-update: $z = z \rho(s Gc)$.
- r-update: $r = r \gamma (r Dc)$.

end while

4. EXPERIMENTAL RESULTS

4.1. Convergence to Ideal MLE Solution

The purpose of the first experiment is to verify the convergence of the proposed algorithm (without the TV penalty) towards the theoretical MLE solution. To this end, we consider the case where the photon count s_m is piecewise constant within each subinterval $Kn \leq m \leq K(n+1)$ for $n=0,\ldots,N-1$. This implies that the discrete filter g_m is a box function of support [0,K]. Consequently, the MLE has a closed form solution [10]

$$\boldsymbol{c}^{(\text{MLE})} = \mathcal{P}_{\mathcal{D}} \left(-\log \left(1 - \boldsymbol{G}^T \boldsymbol{b} / K^2 \right) \right),$$
 (29)

where $\mathcal{P}_{\mathcal{D}}$ is the projection onto the search domain $\mathcal{D} \stackrel{\text{def}}{=} [0,1]^N$.

As a comparison we also consider a gradient descent algorithm presented in [15]. The gradient descent algorithm is applied to solve a simplified case of (11) where there is no total variation penalty (i.e., $\lambda=0$). In this case, we substitute the constraint s=Gc into the objective of (11) to obtain an unconstrained minimization problem in c. We expect the gradient descent iterates to converge to $c^{(\mathrm{MLE})}$ because the objective F(c) of the unconstrained problem is convex. At the kth iteration, the descent update is given by

$$c^{(k+1)} = c^{(k)} - \alpha \nabla F(c^{(k)}) / ||\nabla F(c^{(k)})||,$$
 (30)

where the gradient is (with elementwise multiplication and division)

$$\nabla F(\mathbf{c}) = \mathbf{G}^{T} \left((1 - \mathbf{b}) + \mathbf{b} \cdot \left(\frac{-e^{-\mathbf{G}\mathbf{c}}}{1 - e^{-\mathbf{G}\mathbf{c}}} \right) \right), \quad (31)$$

and the step size α is determined by the standard line search.

Figure 3 shows the mean squared error (MSE) as a function of runtime. It is evident from the figure that the proposed ADMM algorithm has a significantly faster convergence than the descent algorithm. Here, runtime (instead of iteration counts) is considered because the cost per iteration of the descent algorithm is more than that of the ADMM algorithm due to the line search, although the former requires fewer iterations.

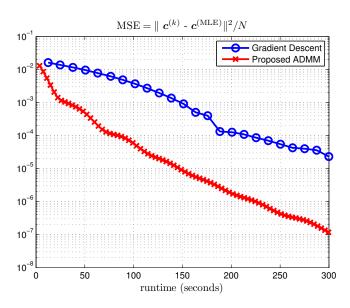


Fig. 3: MSE $\stackrel{\text{def}}{=} \| \boldsymbol{c}^{(k)} - \boldsymbol{c}^{(\text{MLE})} \|^2 / N$ as a function of runtime (sec).

4.2. Reconstruction Quality

To compare the quality of the reconstructed images, we consider an image of size 100×100 . To simulate a realistic scenario, we define the interpolation kernel $\varphi(x)$ as a Gaussian kernel of size 9×9 and unit variance. The oversampling ratio is K=16, and $\beta(x)$ is a box function of size $K \times K$. Poisson measurements are generated and quantized at a threshold of 1 photon count. The observed binary data is fed into the proposed algorithm, and results are obtained when the algorithm terminates.

The internal parameters of the algorithm are defined as follows. We set $\lambda=10$ to put moderate strength of the total variation penalty. The quadratic penalty parameters are set to $\rho=10$ and $\gamma=1$. We stop the iteration of the algorithm when either the number of iterations exceeds a maximum of 100, or the relative change satisfies $\|\boldsymbol{c}^{(k+1)}-\boldsymbol{c}^{(k)}\|/\|\boldsymbol{c}^{(k)}\| \leq 10^{-3}$.

Figure 4 shows the results of the reconstruction. In Figure 4(a), we show the reconstructed result by running the proposed algorithm without total variation (i.e., $\lambda=0$). As illustrated in the previous experiment, this result converges to the ideal MLE solution as we increase the iteration number. For this particular image, the MLE solution achieves a PSNR of 25.62 dB. In Figure 4(b), we show the result when $\lambda=10$. This corresponds to the MAP solution with a total variation prior. Evidently, the PSNR value increases to 30.79 dB, which is significantly higher than that of the MLE solution. In addition, we remark from Figure 3 that the runtime of the proposed ADMM algorithm is much shorter than that of the descent algorithm. These results indicate that the proposed scheme achieves better reconstruction quality with less computation.



(a) MLE Solution [15] $\lambda = 0$ PSNR = 25.62 dB



(b) MAP Solution (proposed) $\lambda = 10$ ${\rm PSNR} = 30.79~{\rm dB}$

Fig. 4: Comparison of the proposed algorithm when total variation penalty is switched on/off. In both cases, we set K=16.

5. CONCLUSION

A fast image reconstruction algorithm for quantum image sensors is proposed. The new algorithm is based on the concept of alternating direction method of multipliers (ADMM). We presented fast solutions to each subproblem associated with the ADMM algorithm. These include a polyphase-domain filtering to achieve efficient matrix inversion, and a nonlinear mapping to find solutions of a transcendental equation. The new algorithm demonstrates superior performance and better convergence speed than the standard gradient descent algorithm. Future work will be focused on the analysis of solutions in the large-scale limits.

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